

**AMENDMENTS TO THE SPECIFICATION**

Please replace paragraph [0037] with the following amended paragraph:

5 Please refer to Fig. 7 (also refer to Fig. 5 and 6). Fig. 7 is a diagram  
of the circuit of the control logic 62 according to the present invention.  
Due to the process of generating the control code simplified by the  
present invention, the control logic 62 only needs inverters 84 and AND  
gates 86. As mentioned above, when the input code 76 represents a  
10 positive value, the sign code A3 is "0", the positive control code 78A  
being equal to the value code 82, all negative control bits X0 to X2  
being "0". According to the control logic 62, the positive control bits Y0  
to Y2 are generated by inverting AND operations of the value sign codes  
82 A0 to A2 and the inverted sign code A3 respectively. When the input  
15 code 76 represents a negative code, the sign code A3 is "1", the negative  
control bits are generated by the inverter of the value code 82, all  
positive control bits being "0". First, the bits A0 to A2 are separately  
converted and are then calculated with the sign code A3 for generating  
the negative control bits X0 to X2. When the sign code A3 is "1", the  
20 control bits X0 to X2 are respectively equal to the inverted values of the  
bits A0 to A2 to define the relationship of Fig. 6. In Fig. 7, the control  
logic in the present invention only needs forty-four transistors (four  
inverters 84 including eight transistors, and six AND gates including  
thirty-six transistors). Compared to the prior art control logic 12  
25 needing more than ninety transistors, the present invention saves a lot of  
transistors and reduces the layout size of the control logic, and further  
reduces the power waste and delays of the gates.

Please replace paragraph [0039] with the following amended paragraph:

30

Please refer to Fig. 8. Fig. 8 is a diagram of the converter in the  
present invention expanded to N bits. A converter 90 receives an N-bit

input code 106 and provides a corresponding output voltage  $V_o$ . There are  $N$  bits  $A(N-1)$  to  $A(0)$  in the input code 106, the bit  $A(N-1)$  being the most significant bit. The converter 90 comprises a control logic 92 and an electrical module 94. The positive control bits  $Y(0)$  to  $[Y(N-1)]$

5    $Y(N-2)$  and the negative control bits  $X(0)$  to  $[X(N-1)]$   $X(N-2)$  are generated by the control logic 92 corresponding to the input code 106. The electrical module 94 comprises a positive electrical module 96A, a negative electrical module 96B, a negative current source 102 as an assistant electrical module, an OP amp 74, and a resistance R. The

10 positive electrical module 96A comprises  $N-1$  positive current sources 98, the positive current sources 98 providing currents  $(2^0)I$ ,  $(2^1)I$  to  $(2^{(N-2)})I$  by ascending powers of two. The switches between the positive current sources 98 and the node A3 are controlled by the positive control bits  $Y(0)$ ,  $Y(1)$  to  $Y(N-2)$ . In other words, the switch

15 corresponding to the positive current source providing  $(2^n)I$  is controlled by the control bit  $Y(n)$  ( $n$  is from 0 to  $(N-2)$ ). The negative electrical module 96B comprises  $N-1$  negative current sources 100. The switch corresponding to the negative current source providing  $(2^n)I$  is controlled by the control bit  $X(n)$  ( $n$  is from 0 to  $(N-2)$ ). The switch 72

20 between the negative current source 102 of the assistant electrical module and the node N3 is controlled by the sign code  $A(N-1)$  of the input code 106,  $A(N-1)$  being the most significant bit of the input code 106. The total current flowing through the node N3 establishes an output voltage through the resistance R.

25